Z-TRANSFORM PROPERTIES

Solution 7.1

(a) - (iii) (b) - (i) (c) - (ii)





(b) Unit circle

Figure S7.2-2

Shown above is the appropriate vector diagram, which we expand out below:



Let d_1 , d_2 , d_3 and d_4 denote the lengths of vectors (1), (2), (3) and (4) respectively. We wish to determine $\frac{d_4}{d_2}$. We know that vector (1) is at an angle of ω to the horizontal axis and that $d_1 = 1$. Therefore:

$$d_{6} = \cos \omega$$

$$d_{3} = \sin \omega$$

$$d_{5} = \frac{1}{a} - d_{6} = \frac{1}{a} - \cos \omega$$

then

$$d_2^2 = (d_6 - a)^2 + d_3^2 = 1 + a^2 - 2a \cos \omega$$

Also,

$$\bar{d}_{4}^{2} = d_{3}^{2} + d_{5}^{2} = 1 + (\frac{1}{a})^{2} - \frac{2}{a}\cos\omega = (\frac{1}{a})^{2}\left[1 + a^{2} - 2a\cos\omega\right]$$

Thus

$$\frac{d_4}{d_2} = \frac{1}{a}$$

Solution 7.3

$$X(z) = \sum_{n=-\infty}^{+\infty} x(n) z^{-n}$$
$$X(\frac{1}{z}) = \sum_{n=-\infty}^{+\infty} x(n) z^{n}$$

With the substitution of variables m = -n in the above summation,

$$X\left(\frac{1}{z}\right) = \sum_{m=-\infty}^{+\infty} x(-m) z^{-m}$$

which we recognize as the z-transform of x(-n).

(ii)

$$X(z) = \sum_{n=-\infty}^{+\infty} x(n) z^{-n}$$

$$\frac{dX(z)}{dz} = \sum_{n=-\infty}^{+\infty} -nx(n) z^{-n-1}$$

$$-z \frac{dX(z)}{dz} = \sum_{n=-\infty}^{+\infty} nx(n) z^{-n}$$

We recognize the right-hand side as the z-transform of n x(n).

Solution 7.4

Letting X(z) and Y(z) denote the z-transforms of x(n) and y(n), and using the properties of z-transforms, the z-transform of the difference equation results in

$$z^{-1} Y(z) - \frac{10}{3} Y(z) + z Y(z) = X(z)$$

thus the system function H(z) is

$$H(z) = \frac{Y(z)}{X(z)} = \frac{1}{z^{-1} - \frac{10}{3} + z} = \frac{z^{-1}}{z^{-2} - \frac{10}{3}z^{-1} + 1}$$
$$= \frac{z^{-1}}{(1 - 3z^{-1})(1 - \frac{1}{3}z^{-1})}$$

To determine the unit-sample response we can obtain the inverse transform of H(z) using any of the methods that we have discussed. For example, using contour integration,

$$\mathbf{x}(\mathbf{n}) = \frac{1}{2\pi j} \oint_{\mathbf{C}} \frac{z^{-1}}{(1 - 3z^{-1})(1 - \frac{1}{3}z^{-1})} z^{\mathbf{n}-1} dz$$

Since the system is stable, the region of convergence includes the unit circle. Thus for $n \geq 0$

$$x(n) = \operatorname{Res}\left[\frac{z^{n}}{(z-3)(z-\frac{1}{3})} \text{ at } z = \frac{1}{3}\right] = -\frac{3}{8}(\frac{1}{3})^{n}$$

For n < 0

$$\mathbf{x}(n) = \frac{1}{2\pi j} \oint_{\mathbf{C}} \frac{\mathbf{p}}{(1 - 3p)(1 - \frac{1}{3}p)} p^{-n-1} dp$$

$$= \operatorname{Res} \left[\frac{p^{-n}}{(1 - \frac{9}{3}p)(1 - \frac{1}{3}p)} \text{ at } p = \frac{1}{3} \right] = -\frac{3}{8} \left(\frac{1}{3}\right)^{-n}$$

therefore,

$$x(n) = -\frac{3}{8}\left[\left(\frac{1}{3}\right)^n u(n) + (3)^n u(-n-1)\right]$$

Solution 7.5

(a) According to the differentiation property, the z-transform of nx(n)is $-z \frac{dX(z)}{dz}$. For this problem,

$$-z \frac{dX(z)}{dz} = - \frac{az^{-1}}{(1 - az^{-1})}$$

The inverse z-transform of $-\frac{a}{(1-az^{-1})}$ is $-a(a)^n u(n)$. From the shifting property, then, the inverse z-transform of $\frac{-az^{-1}}{(1-az^{-1})}$ is $-a(a)^{n-1}u(n-1)$.

Therefore

$$n \mathbf{x}(n) = -a^{n} \mathbf{u}(n-1)$$

or
$$\mathbf{x}(n) = -\frac{a^{n}}{n} \mathbf{u}(n-1) \qquad n \neq 0$$

Note that since we obtain nx(n) from the differentiation property, this does not allow us to obtain x(0). For this problem, however, we could obtain x(0) from problem 5.7. Specifically, since x(n) is causal,

$$x(0) = \lim_{z \to \infty} \log(1 - az^{-1}) = 0$$

(b) For
$$|\rho| < 1$$
, the power series expansion for log $(1 - \rho)$ is
log $(1 - \rho) = -\sum_{n=1}^{\infty} \frac{\rho^n}{n}$

thus,

log(l -
$$az^{-1}$$
) = - $\sum_{n=1}^{\infty} \frac{a^n}{n} z^{-n}$
thus we identify x(n) as
x(n) = - $\frac{a^n}{n}$ u(n - 1)

Note this problem is strongly related to the discussion in chapter 12 of the text. You may wish to look through some of the discussion in that chapter.

Solution 7.6

$$X_{1}(z) = \sum_{n=-\infty}^{+\infty} x_{1}(n) z^{-n} = \sum_{n=-\infty}^{+\infty} x(n) z^{-nM}$$
Therefore $X_{1}(z) = X(z^{M})$
For $M = 2$
 $X_{1}(e^{j\omega}) = X(e^{j2\omega})$





Resource: Digital Signal Processing Prof. Alan V. Oppenheim

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